

Noise in Short Channel MOSFETs

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Overview

- **Creativity in Analog / Mixed Signal IC Design**
- **DSM CMOS Effects on Analog Design**
- **Fundamental Noise Sources**
- **Applications**
- **Conclusion**

Overview

- **Creativity in Analog / Mixed Signal IC Design**
 - **Role of Creativity**
- DSM CMOS Effects on Analog Design
- Fundamental Noise Sources
- Applications
- Conclusion

Need for Creativity:

"... every company is betting on the ingenuity of its engineers to create the products that will conquer the market. And yet engineers' conferences and travels are cut, as if they don't make any contribution to creativity - the creativity that is required, ironically, at the very moment that its sources of inspiration are being cut."

**- Willy Sansen,
IEEE SSCS Magazine,
Summer 2009, p.4**

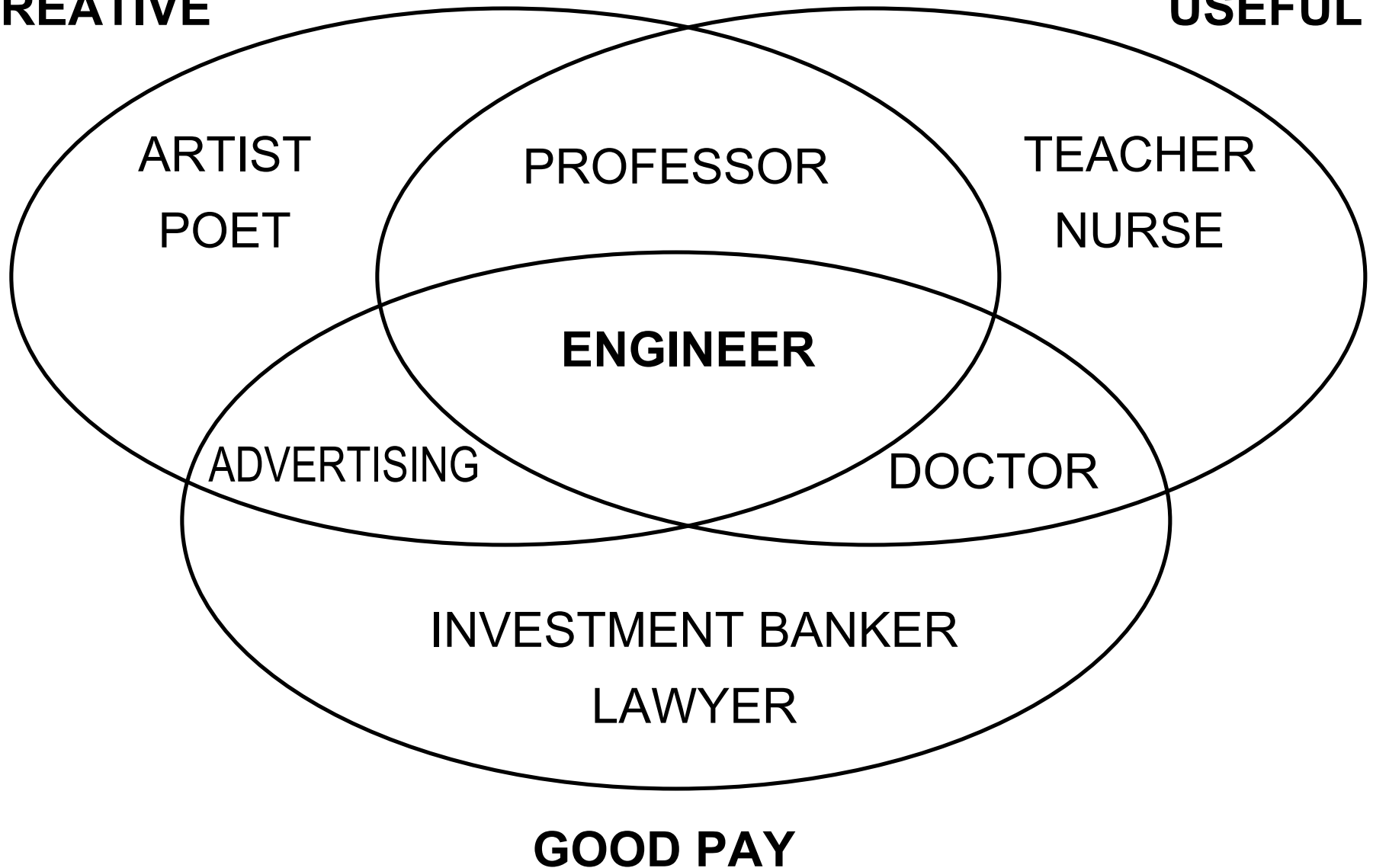
Why be creative?

- **Need**
 - **Easy problems solved already**
 - **Tough problems need creative solution**
- **Dealing with environment of change**
 - **Coping vs. thriving**
- **Human nature**
 - **Fun!**

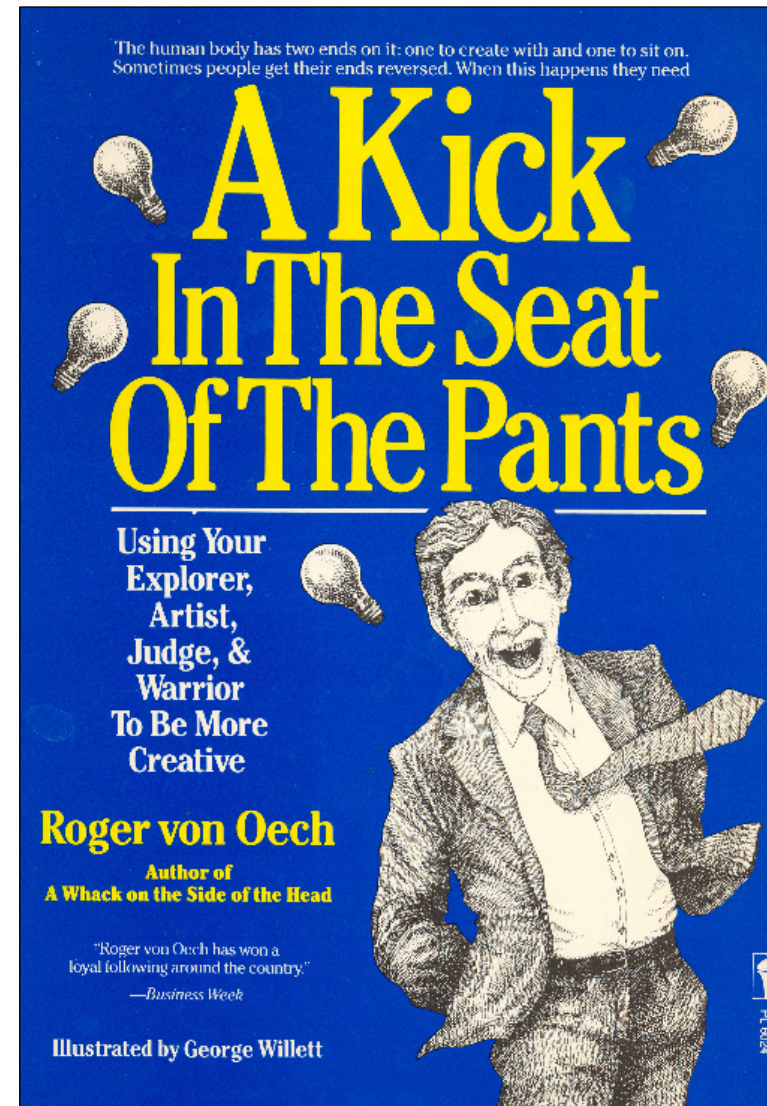
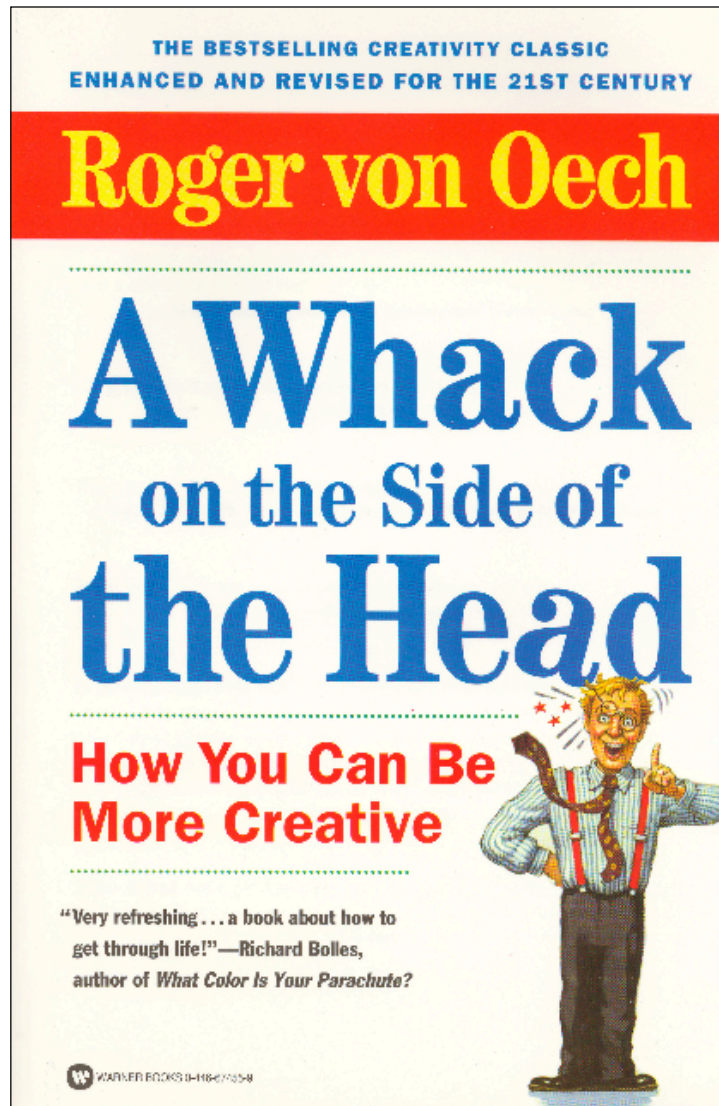
Career Classification

CREATIVE

USEFUL



Creativity Resources



Creativity Framework

Explorer

Artist

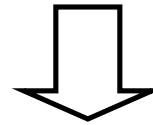
Judge

Warrior

Example: Time (Stages of project)

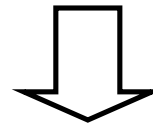
Explorer

Background Research



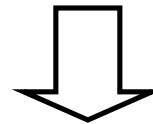
Artist

Brainstorm Options



Judge

Choose Solution



Warrior

Implement Design

Creativity Framework

Explorer

Seek out new information

Survey the landscape

Get off the beaten path

Artist

Poke around in unrelated areas

Gather lots of ideas

Shift your mindset

Judge

Don't overlook the obvious

Look for unusual patterns

Warrior

Creativity Framework

Explorer

Create something original

Multiply options

Use your imagination

Artist

Ask "what if" questions

Play with ideas

Look for hidden analogies

Judge

Break the rules

Look at things backward

Change contexts

Warrior

Play the fool

Creativity Framework

Explorer

**Evaluate options
Ask what's wrong
Weigh the risk**

Artist

**Embrace failure
Question assumptions
Look for hidden bias**

Judge

**Balance reason and hunches
Make a decision!**

Warrior

Creativity Framework

Explorer

**Put decision into practice
Commit to a realistic plan
Get help**

Artist

**Find your real motivation
See difficulty as challenge
Avoid excuses**

Judge

**Persist through criticism
Sell benefits not features
Make it happen**

Warrior

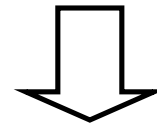
Learn from every outcome

Example: Modes of Thinking

Explorer

Divergent
Soft
Qualitative

Artist



Judge

Convergent
Hard
Quantitative

Warrior

Why a Creativity Model?

Education

- **Standardized-test-numbered students**
- **Paralysis in face of open-ended problem**

Designer

- **Awareness of strengths, weaknesses**
- **Recognize preferences**

Not Right or Wrong!

- **One way of looking at process**
- **Orchard analogy**

Creativity Framework

Explorer

Survey the landscape

Artist

Break the rules

Judge

Question assumptions

Warrior

Learn from every outcome

Overview

- Creativity in Analog / Mixed Signal IC Design
- **DSM CMOS Effects on Analog Design**
 - Short Channel Effects
 - Noise Behavior
- Fundamental Noise Sources
- Applications
- Conclusion

Survey the landscape

Good Old Days

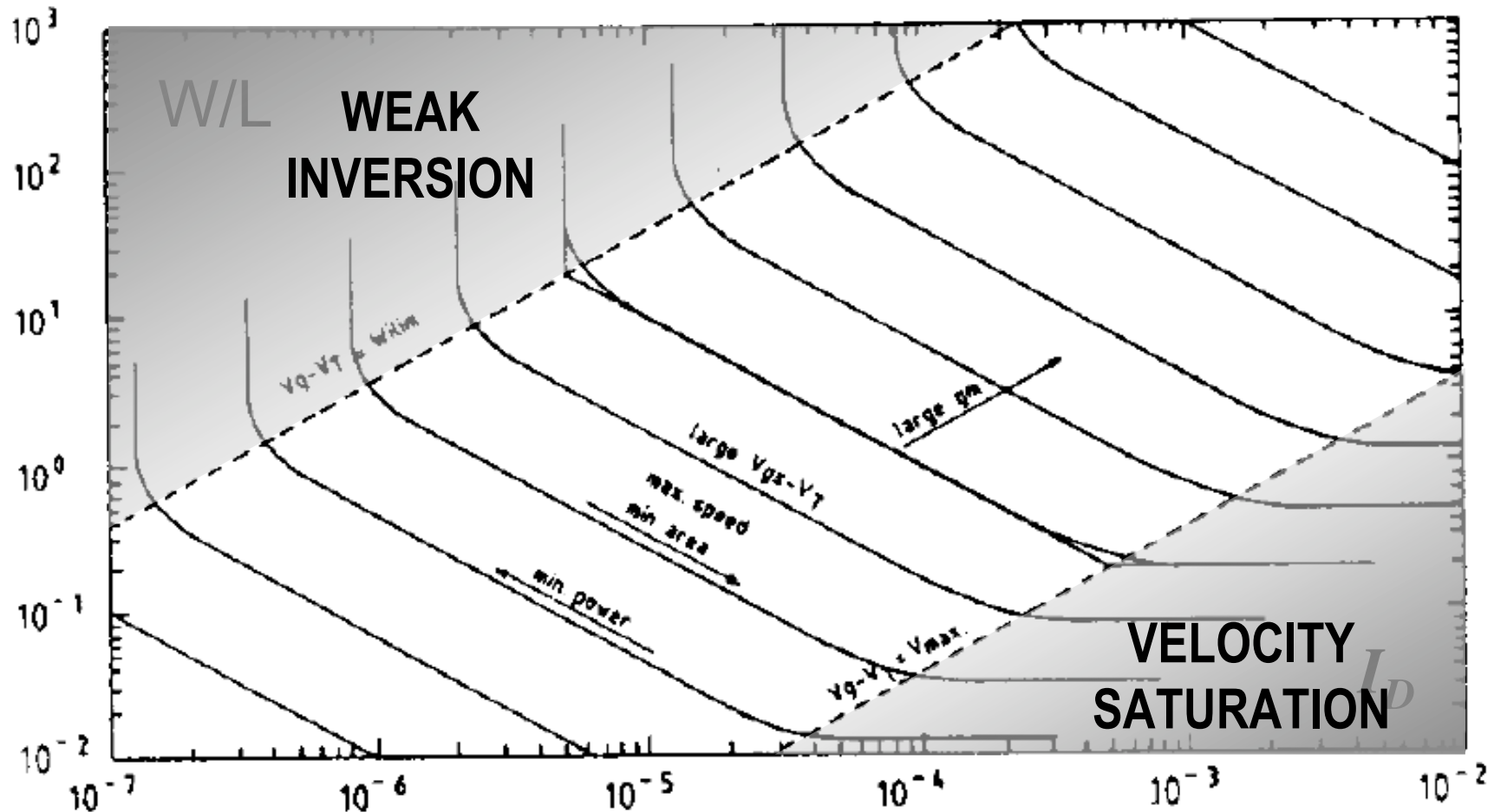
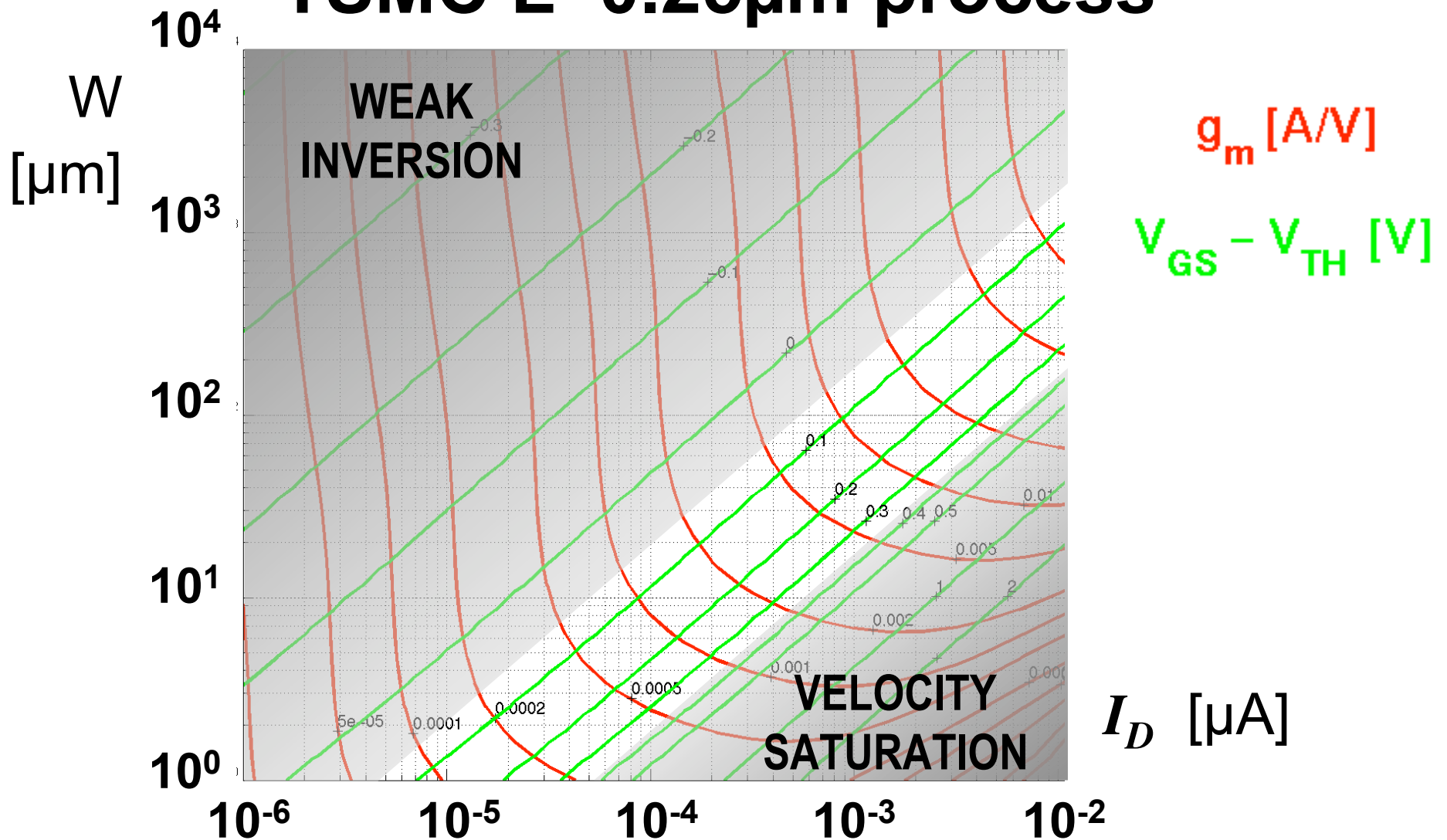


Fig.1: Curves of constant g_m in the I - W/L plane

- **Large strong inversion region**
 - **"Square law", easy hand analysis**

TSMC L=0.25 μm process



- ~~Square law~~
- Graphical / numerical analysis

MOSFET Noise

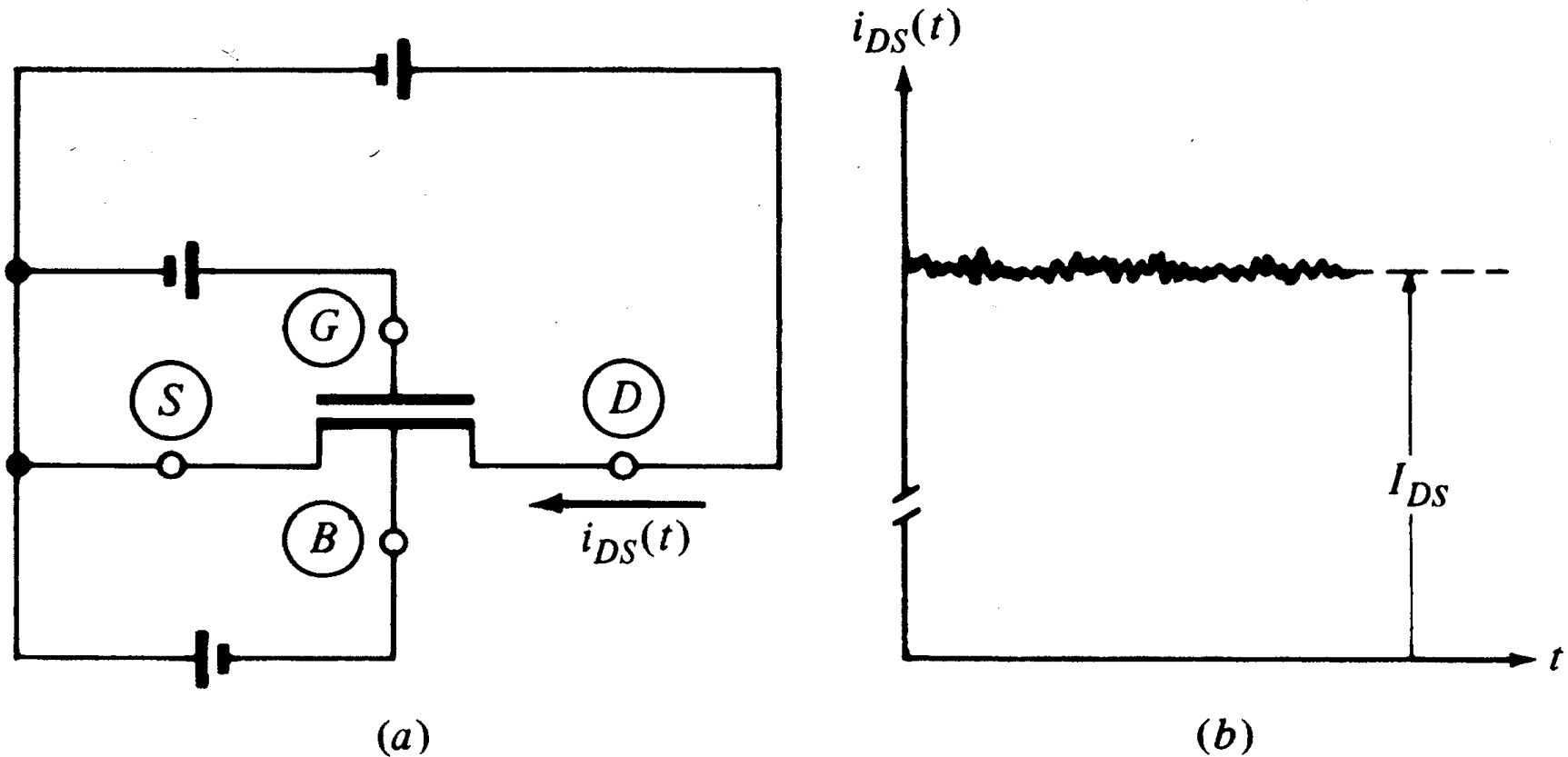


FIGURE 8.31

- (a) A MOS transistor biased with fixed noiseless terminal voltages;
(b) the drain-to-source current for the connection in (a), including noise.

MOSFET Noise p.s.d.

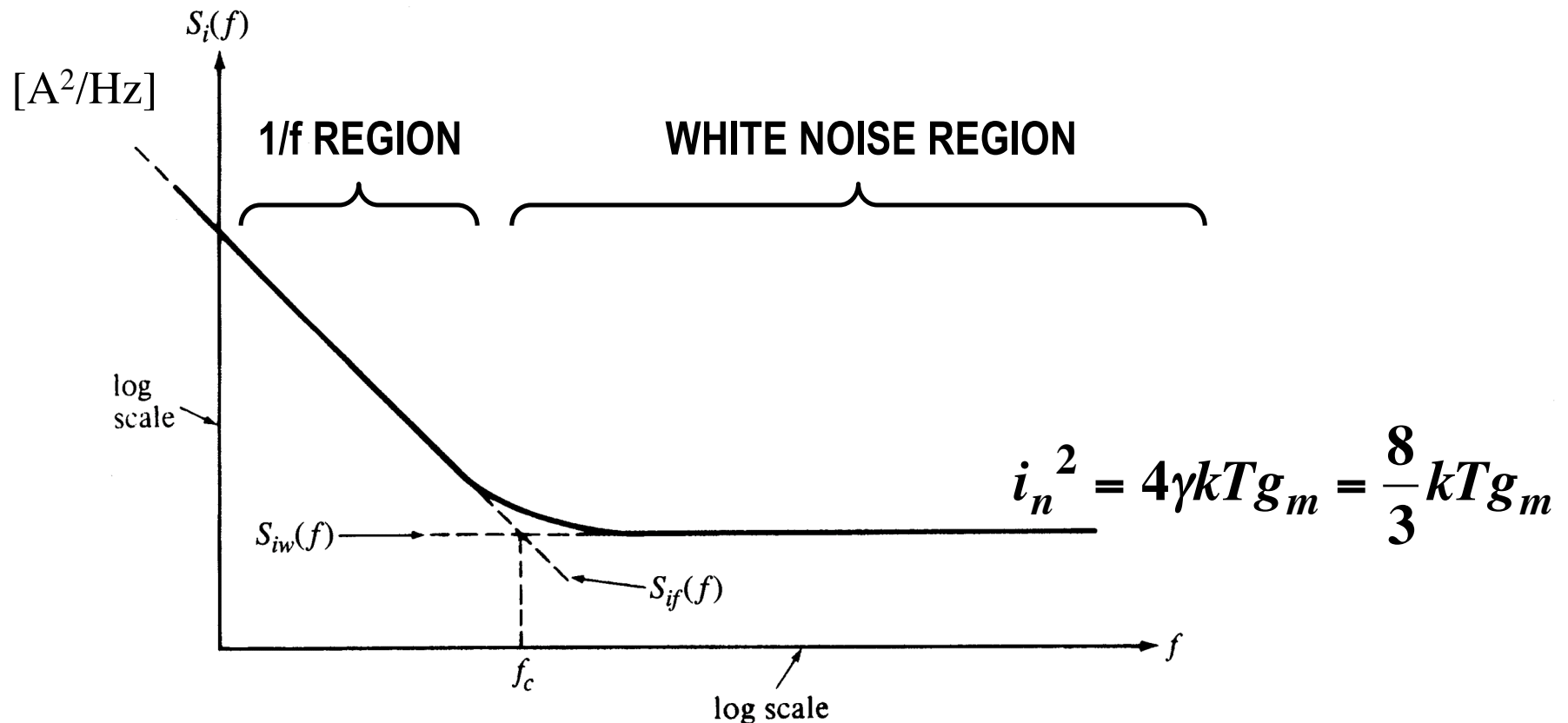


FIGURE 8.33

A typical plot of the drain-noise current power-spectral density versus frequency in log-log axes. Subscript i refer to total noise, i_w to white noise, and i_f to flicker noise.

- **Saturation, strong inversion operation**
- **Where does factor $\gamma=2/3$ come from?**

Submicron CMOS: Noise behavior

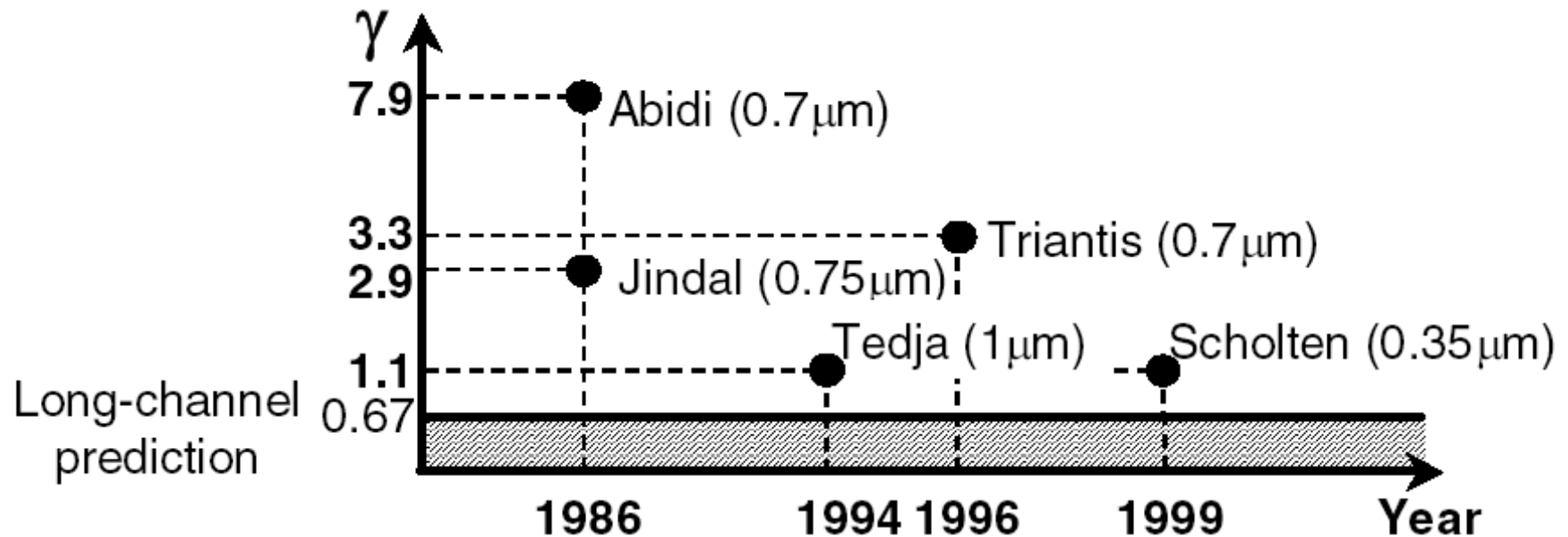


Fig. 1. Some reported values of Gamma noise factor in various technologies.

⇒ **Gamma factor $\gamma > 2/3$?!?**

Disagreement with long channel model?

Question assumptions

Overview

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 - **Shot Noise**
 - **Thermal Noise**
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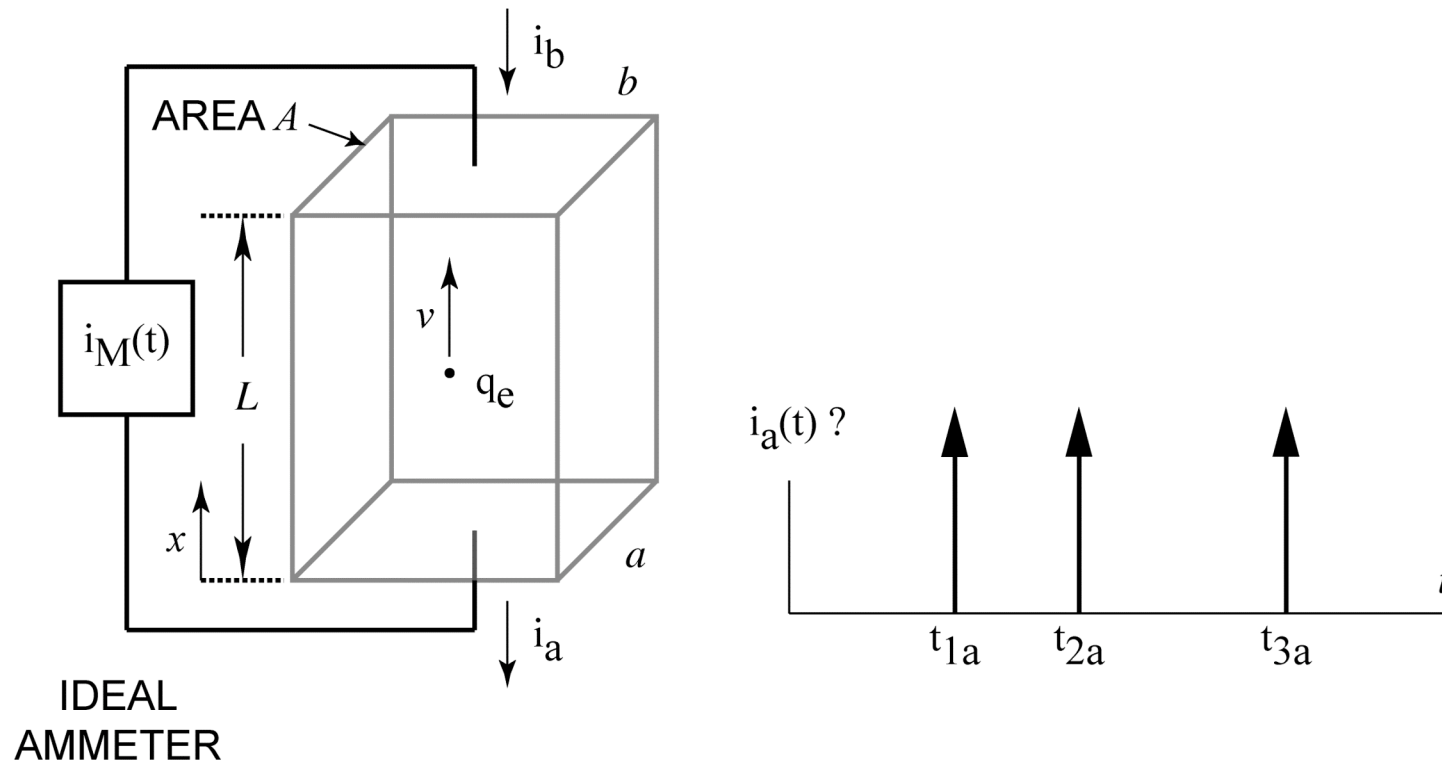
Shot Noise

- Current noise density for DC current I_{DC}

$$i_n^2 = 2q_e I_{DC}$$

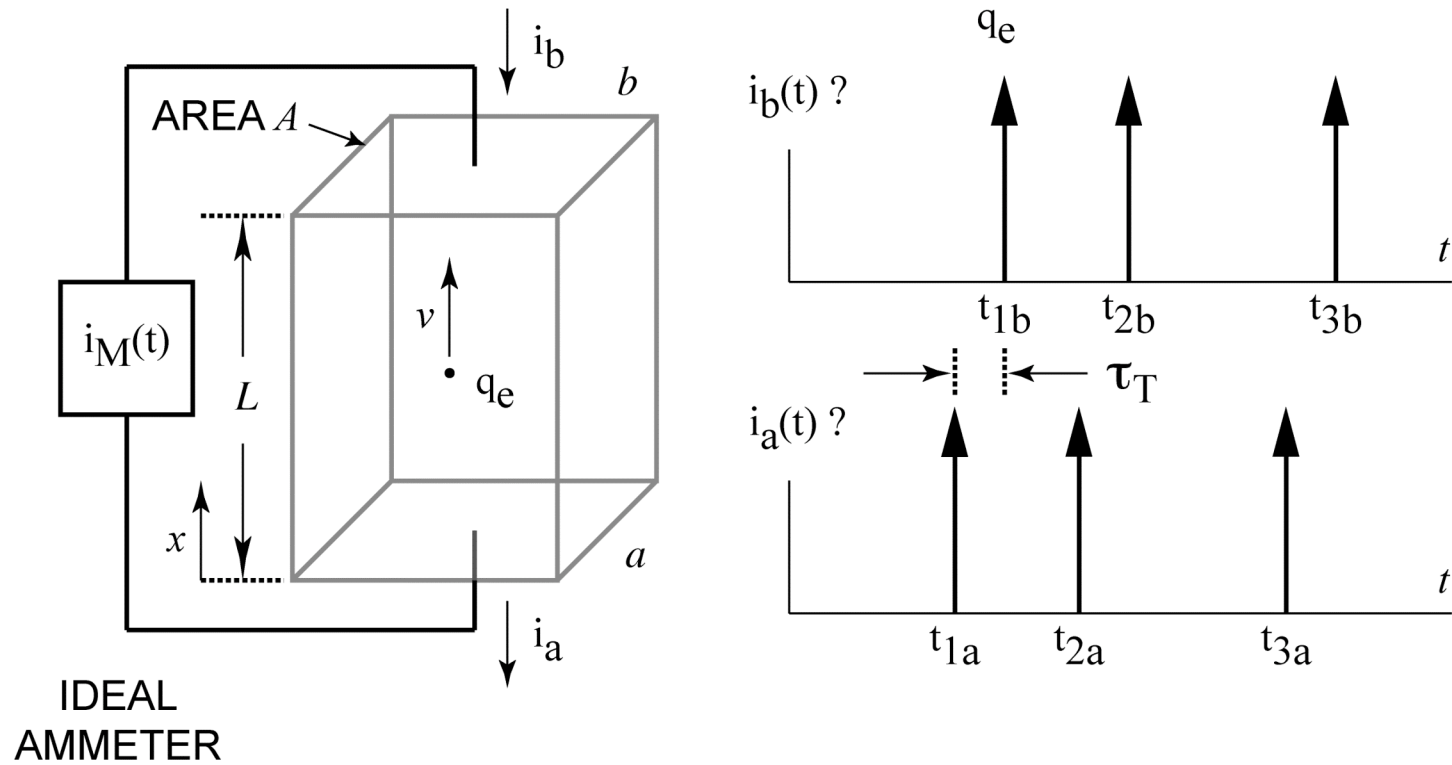
- Where does this come from?
- Key assumption:
 - Electron arrivals independent events

Shot Noise



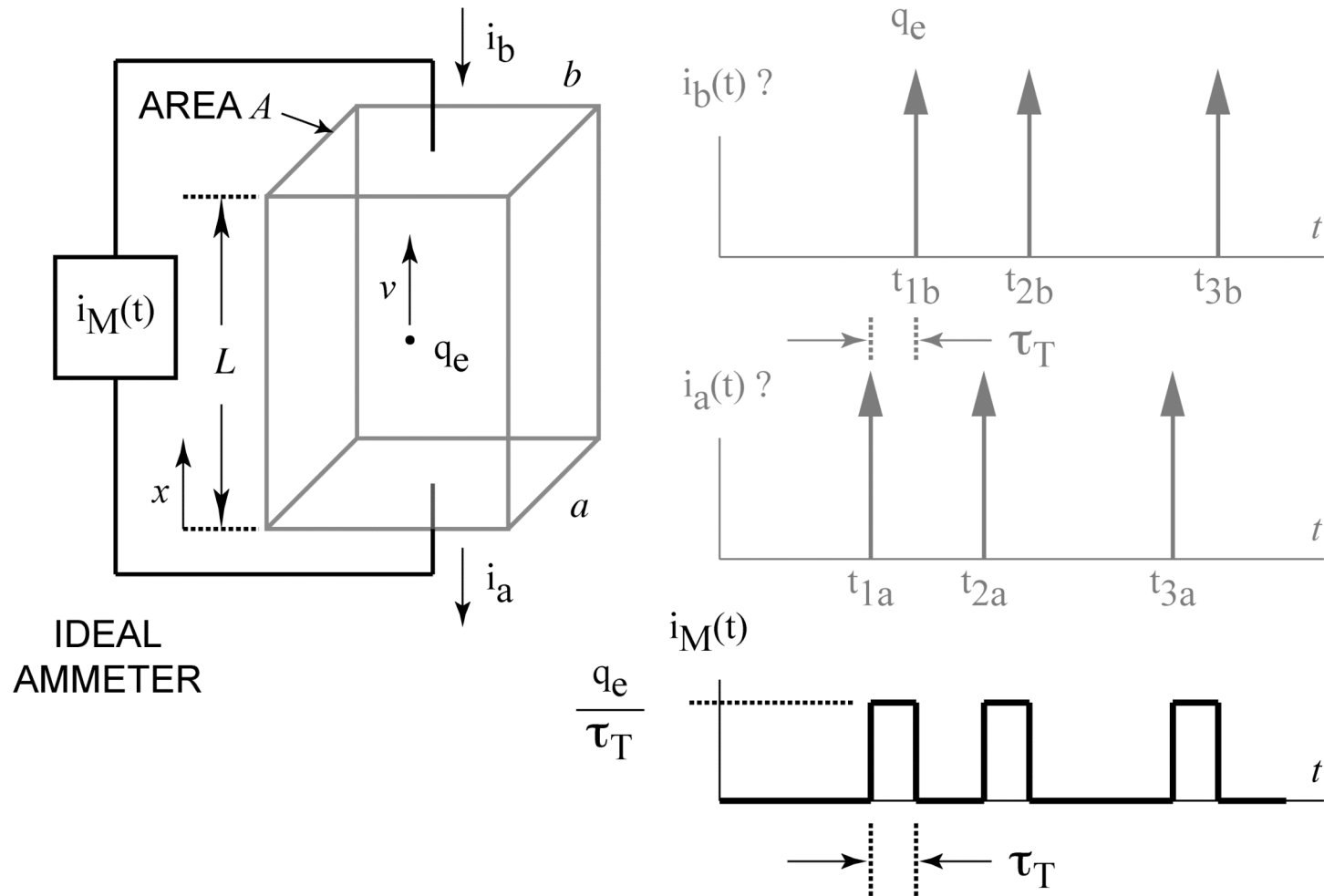
- What is current measured by ammeter?

Shot Noise



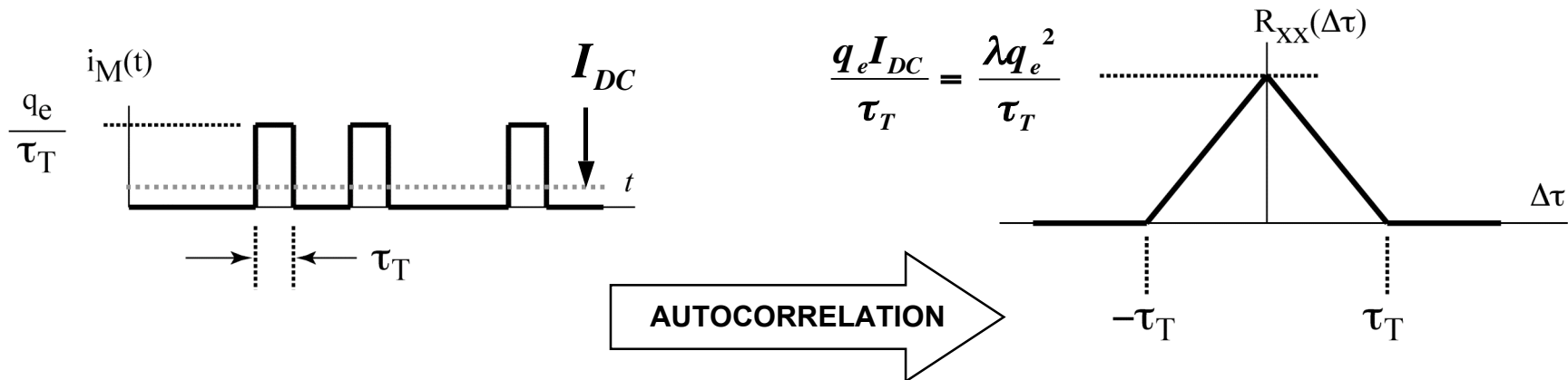
- What is current measured by ammeter?

Ramo-Shockley Theorem



- **Current measured by ammeter:**
 - Randomly arriving pulses with area q_e

Poisson Process

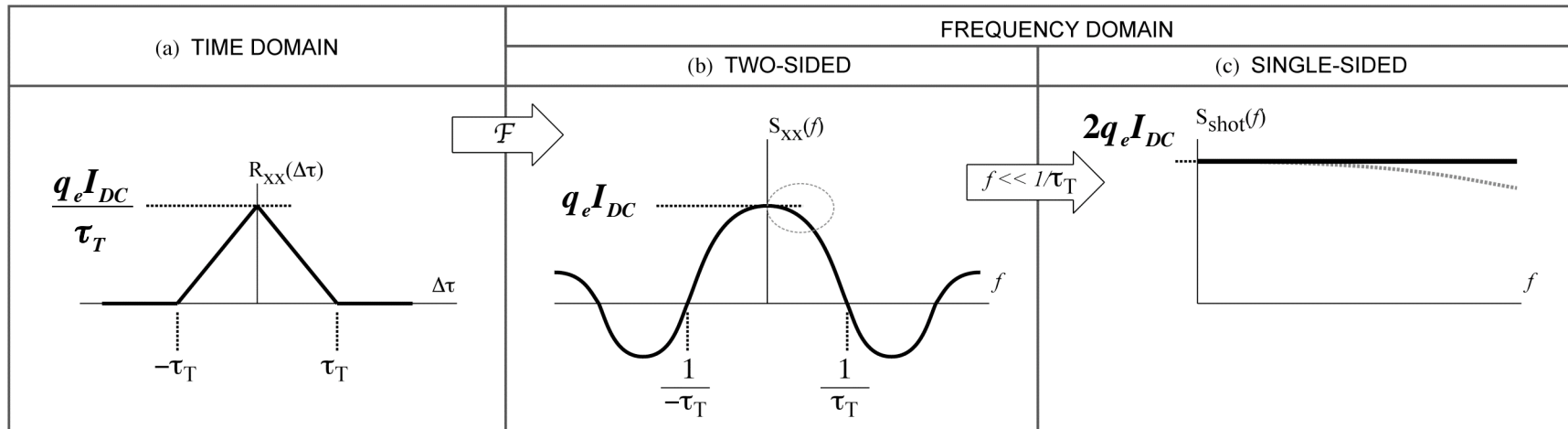


- Average arrival rate λ [sec⁻¹]
- Average DC current:

$$I_{DC} = \lambda q_e$$

- Autocorrelation: time domain description of random process

Shot Noise Power Spectral Density



- **Wiener-Khinchine theorem**
 - Autocorrelation → frequency domain p.s.d
- **Frequency domain**
 - For frequencies $< 1/\tau_T$

$$i_n^2 = 2q_e I_{DC}$$

Shot Noise Power Spectral Density

- **Key Points:**
 - Discrete nature of charge is essential
 - Carrier transits are independent events
 - Carriers do not interact with each other or with any medium
 - Temperature not a factor

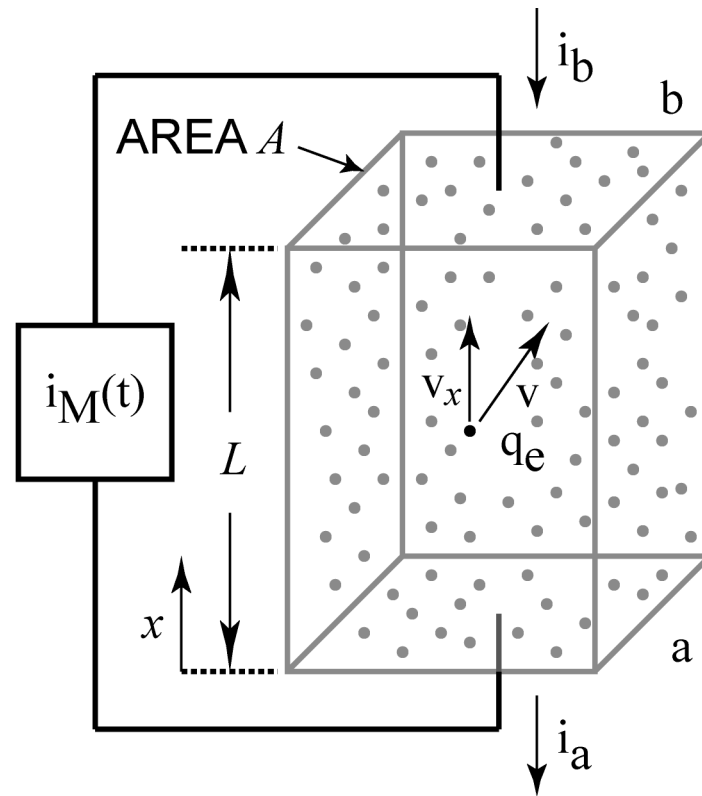
Thermal Noise

- **Current noise density for resistor**

$$i_n^2 = \frac{4kT}{R}$$

- **Where does this come from?**
- **Assumption:**
 - **Carriers in thermal equilibrium**

Thermal Noise in Resistor



- **Assumption:**
 - Carriers in thermal equilibrium
- Random velocity vectors \mathbf{v}
- Only v_x component contributes to current

Boltzmann's Constant k

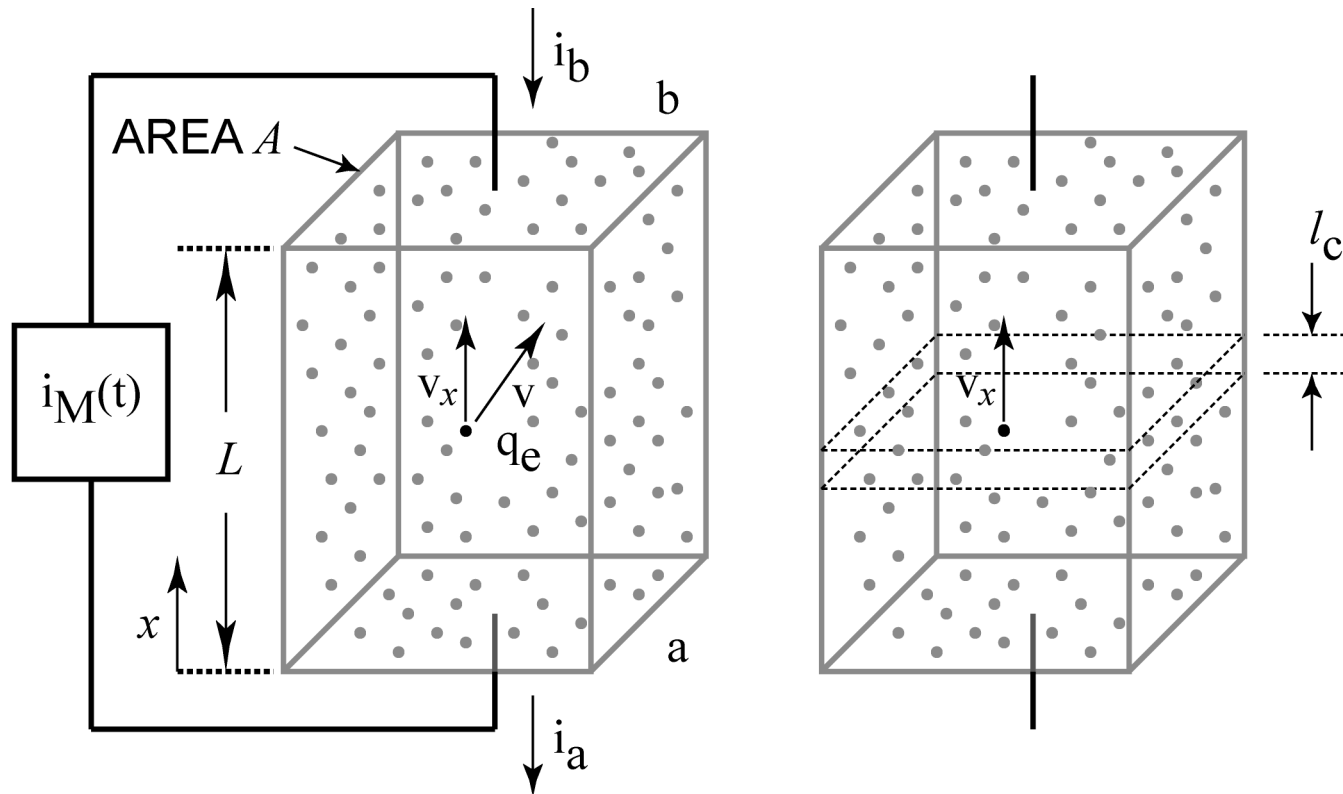
- $k = 1.38 \text{ E-23 J/K}$ Meaning?
- Thermodynamics: Equipartition theorem
 - Independent energy storage modes in a system at equilibrium have average energy of $kT/2$
 - Equivalent statements:

"Temperature in this room is 293K"

"Average kinetic energy (in each of x, y, z directions) for each air molecule in this room is 2.02E-21 joule"

$$\frac{kT}{2} = \frac{(1.38 \text{E} - 23 [\text{J/K}])(293 [\text{K}])}{2} = 2.02 \text{E} - 21 [\text{J}]$$

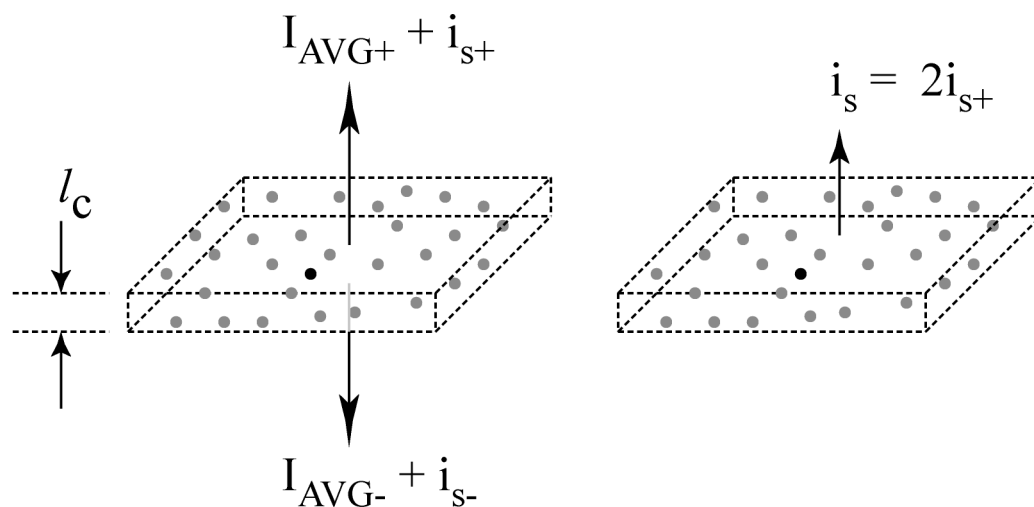
Thermal Noise



- **Approximate collision statistics:**

Mean free path	l_c	0.1 μm
Mean free time	τ_c	1 ps
Velocity (rms)	v_x	0.1 $\mu\text{m/ps}$

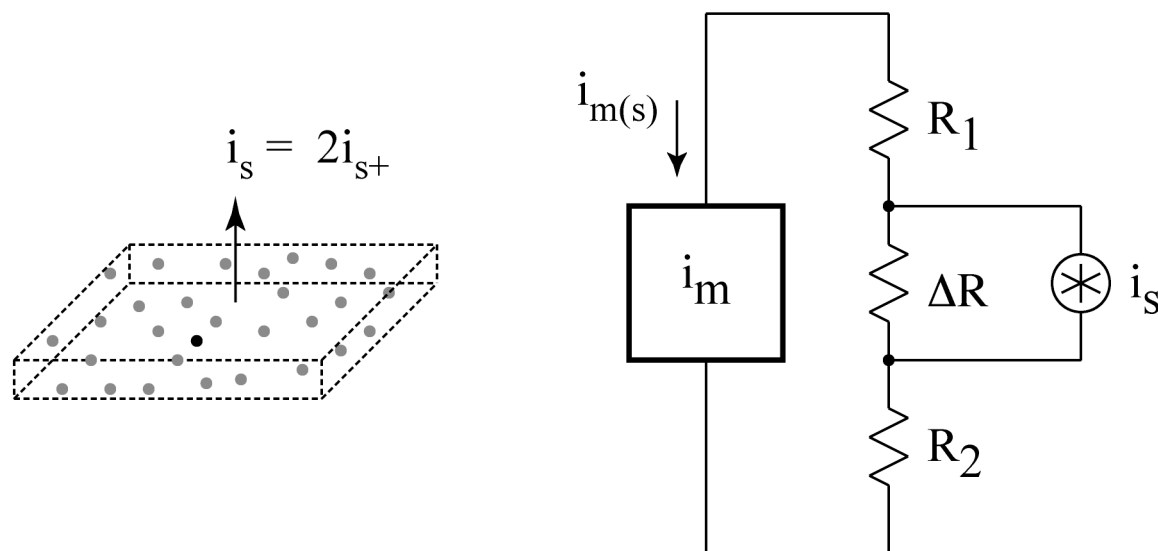
Thermal Noise



- Consider "slice" equal to mean free path l_c
- During one mean free time τ_c
 - On average, half of carriers exit each way:

$$I_{AVG+} = I_{AVG-}$$
- Shot noise components $i_{s+} = -i_{s-}$ correlated
 - Noise current from "slice" $i_s = 2i_{s+}$

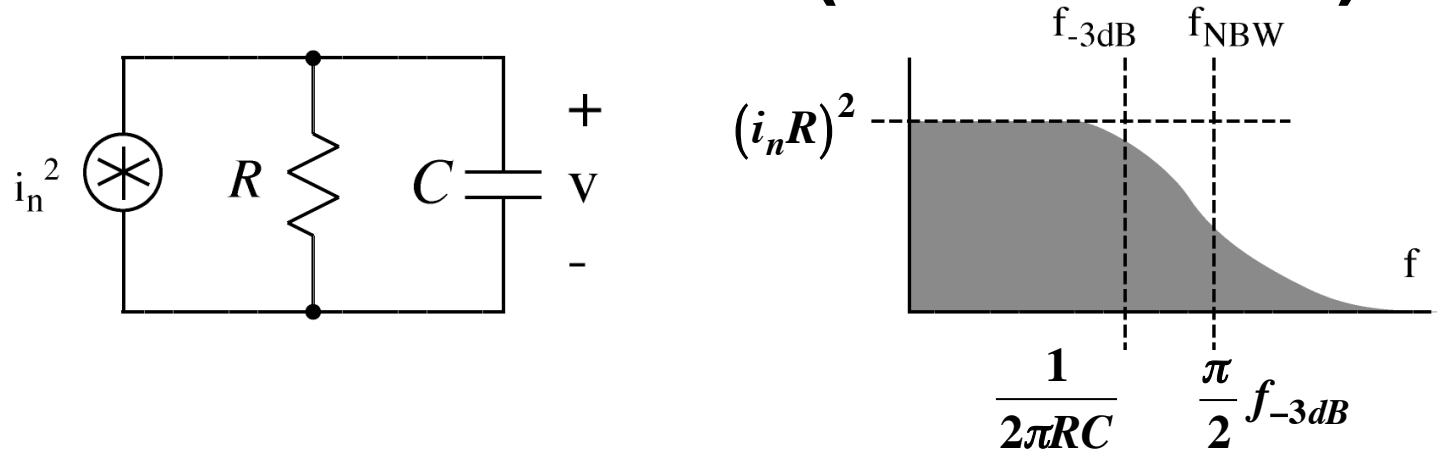
Thermal Noise



- Sum (independent) contributions from slices
- Noise current seen by external ammeter $i_{m(s)}$ reduced by current divider factor:
 ΔR of slice, total resistance $R = R_1 + \Delta R + R_2$
- Relating to R using mobility definition gives

$$i_n^2 = \frac{4kT}{R}$$

Thermal Noise (Alternative)



- Equipartition, rms energy in capacitor:

$$\frac{1}{2} C v^2 = \frac{kT}{2} \Rightarrow v^2 = \frac{kT}{C}$$

- Integrate noise p.s.d. over noise bandwidth:

$$v^2 = (i_n R)^2 \left(\frac{\pi}{2} \frac{1}{2\pi RC} \right) \Rightarrow v^2 = i_n^2 \frac{1}{4RC}$$

- Equate:

$$\frac{kT}{C} = i_n^2 \frac{1}{4RC} \Rightarrow \boxed{i_n^2 = \frac{4kT}{R}}$$

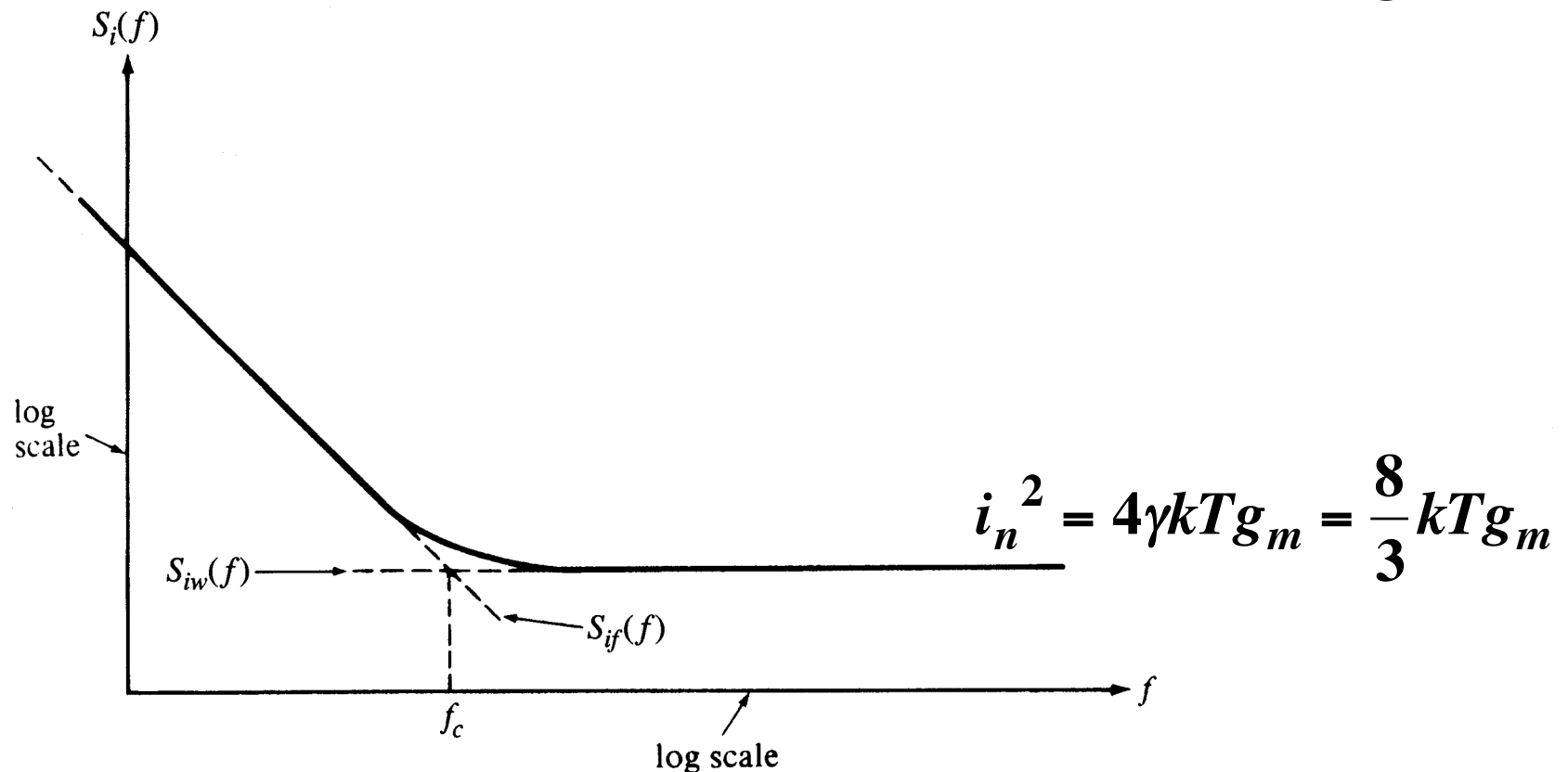
Thermal Noise Power Spectral Density

- **Key Points:**
 - **Discrete nature of charge is not essential**
 - **Can also be derived from equipartition only (e.g. kT/C noise)**
 - **Carrier scattering: interact with medium, thermal equilibrium**
 - **Carrier transits are not independent due to interaction with medium**
 - **Temperature is important to determine carrier average kinetic energy / velocity**

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 - **Oscillator Jitter**
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MOSFET Channel Noise Density



- Where does this come from?
- Assumption:
 - Resistive channel segments

MOSFET Noise Analysis

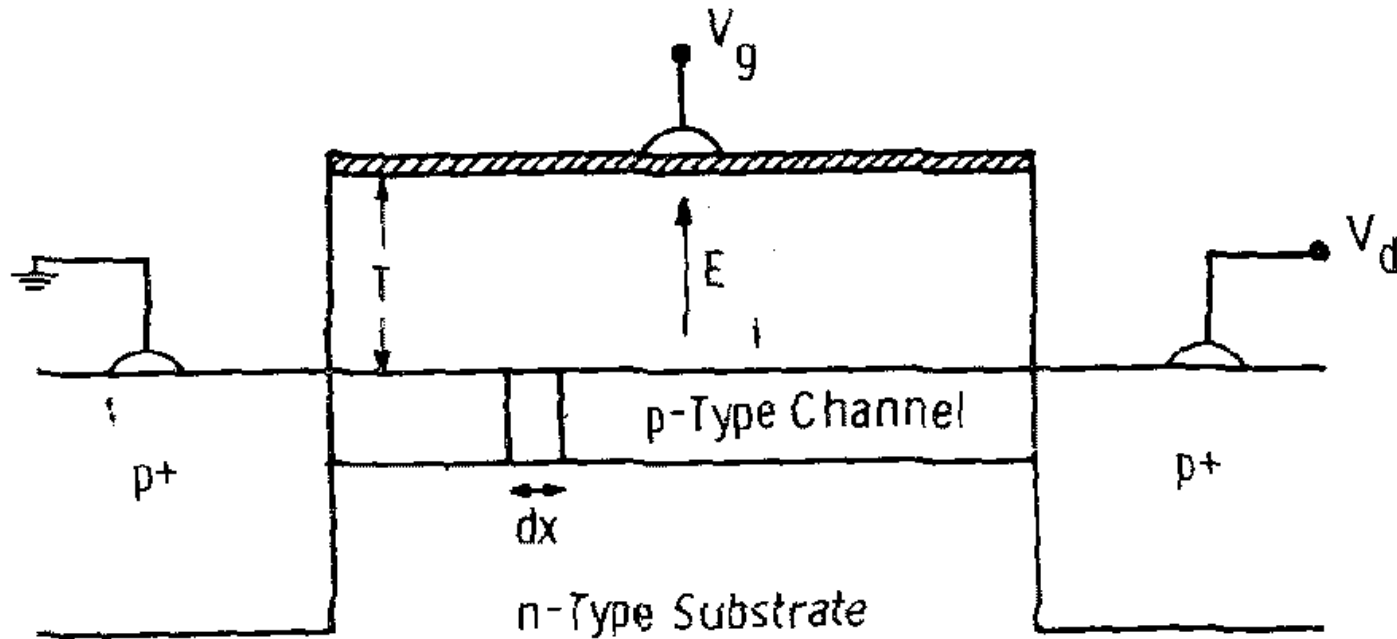


Fig. 6. Schematic representation of MOS transistor.

- **Model: Thermal noise $d\nu$ for differential segment dx of MOSFET channel**
- **Integrate over channel length L**
- **Gamma factor $\gamma = 2/3$ falls out of integral**

MOSFET Noise Analysis

- **Key assumption:**
 - Carrier behavior in channel determined by mobility (resistive) behavior

Ask "what if" questions

- What if it's not a resistor?

Velocity Saturation

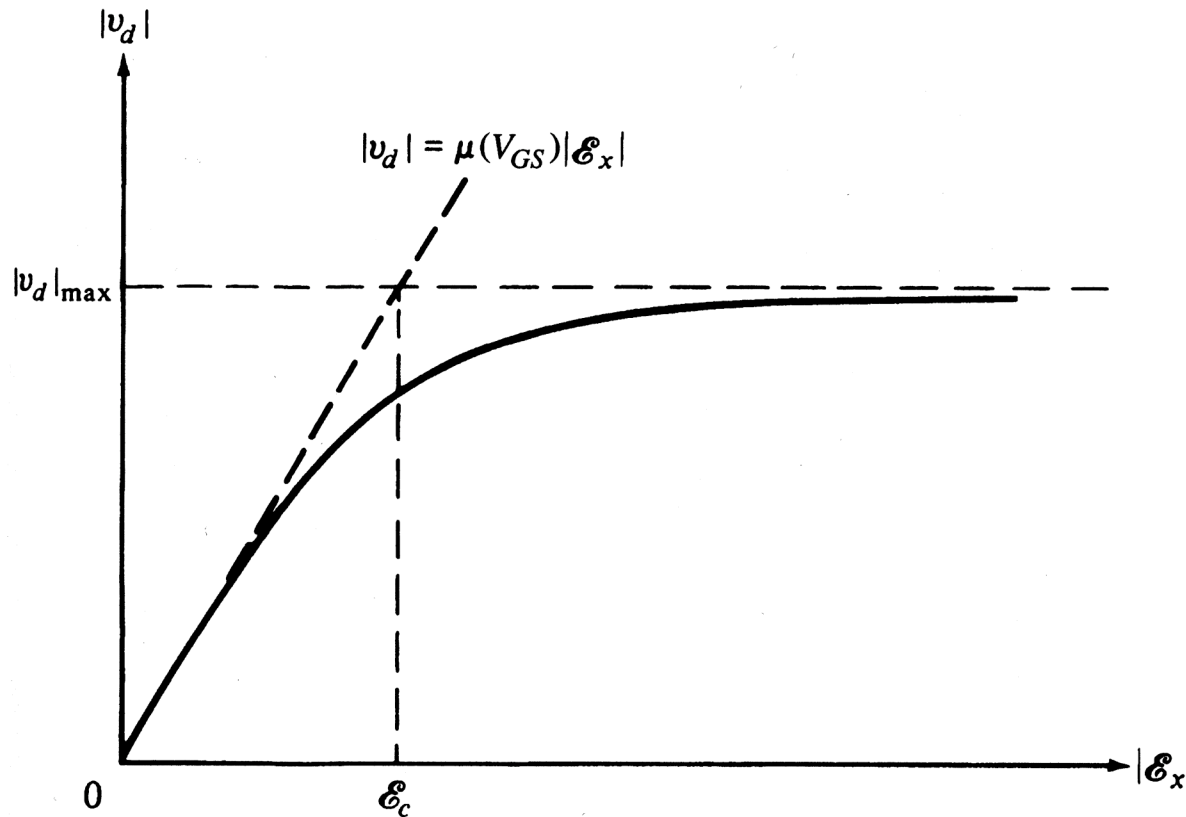
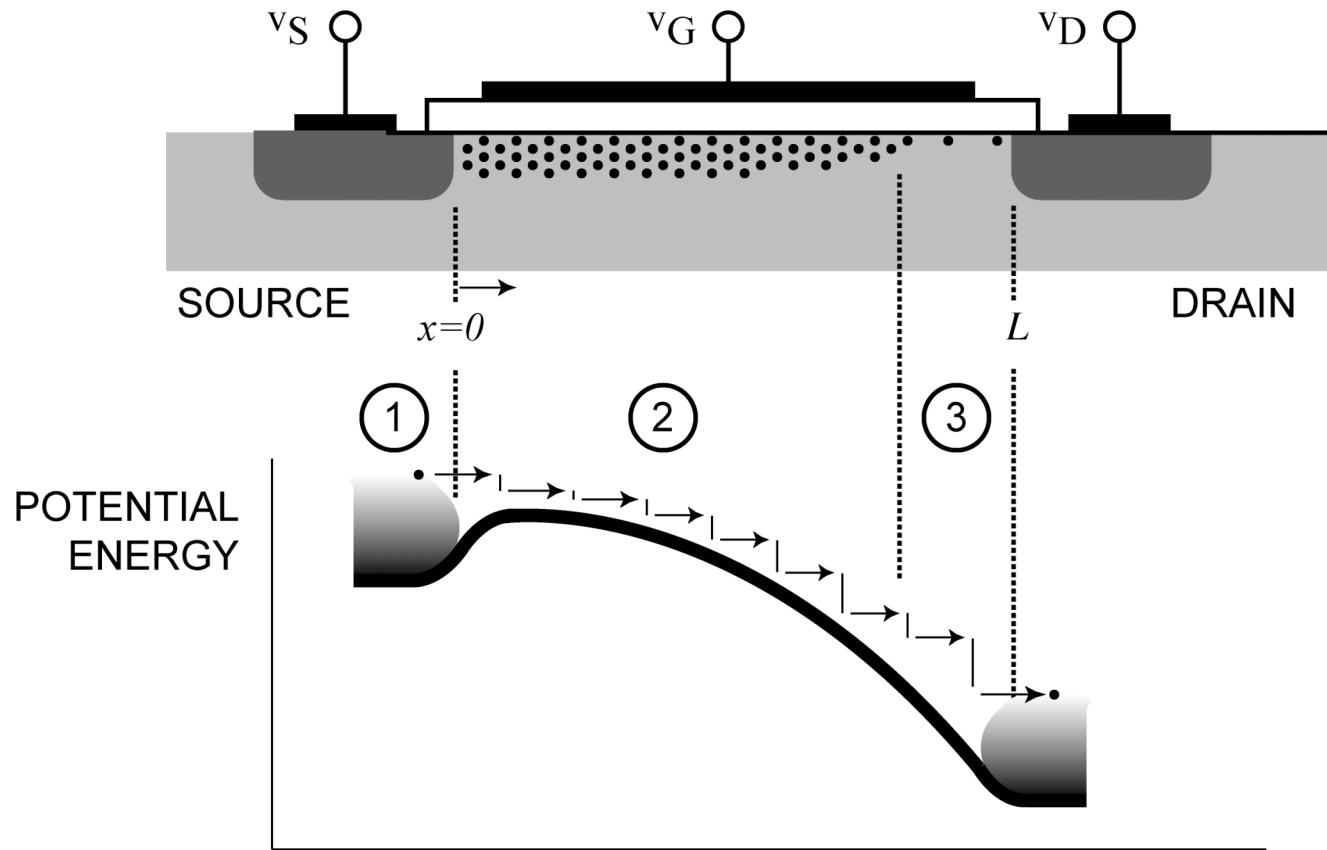


FIGURE 6.22

Magnitude of carrier velocity in the inversion layer vs. magnitude of longitudinal component of electric field, $|\mathcal{E}_x|$. $\mu(V_{GS})$ is the low $|\mathcal{E}_x|$ field surface mobility at a given V_{GS} (see Sec. 4.10).

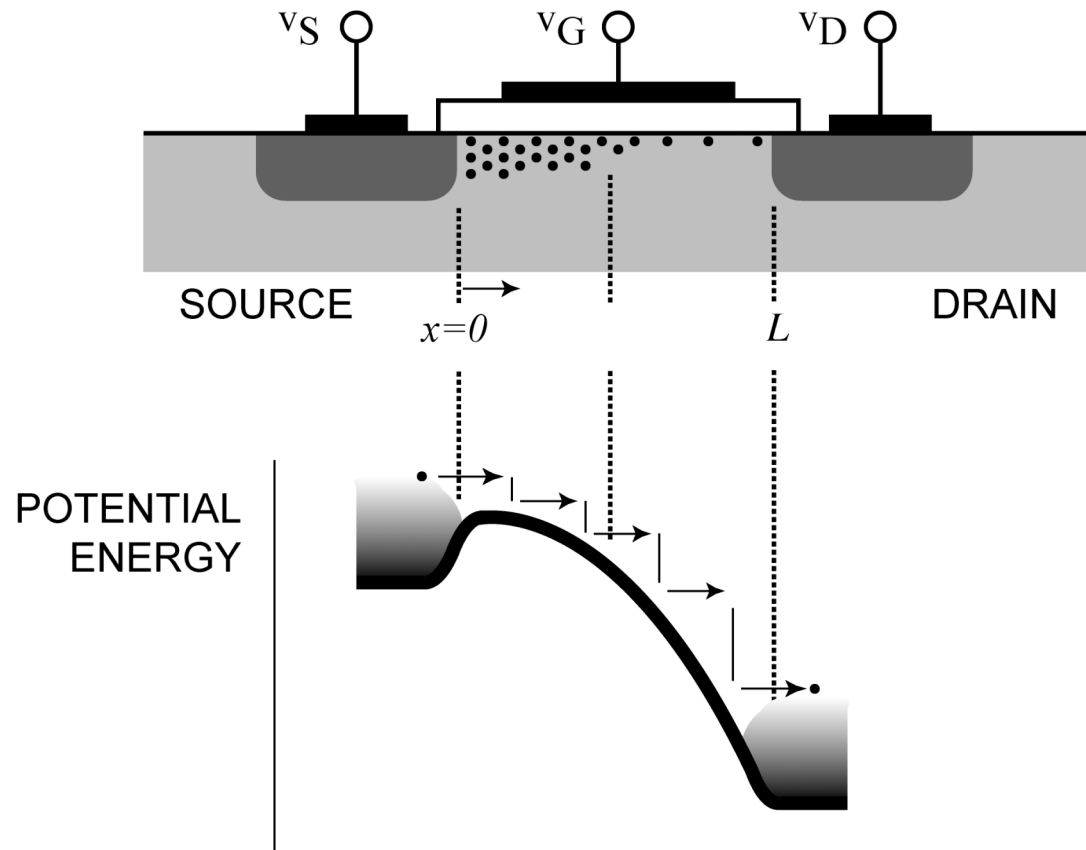
- **Deviation from mobility model at high field**
 - "High field" \Leftrightarrow Small dimensions

MOSFET Potential Energy ($L \sim \mu\text{m}$)



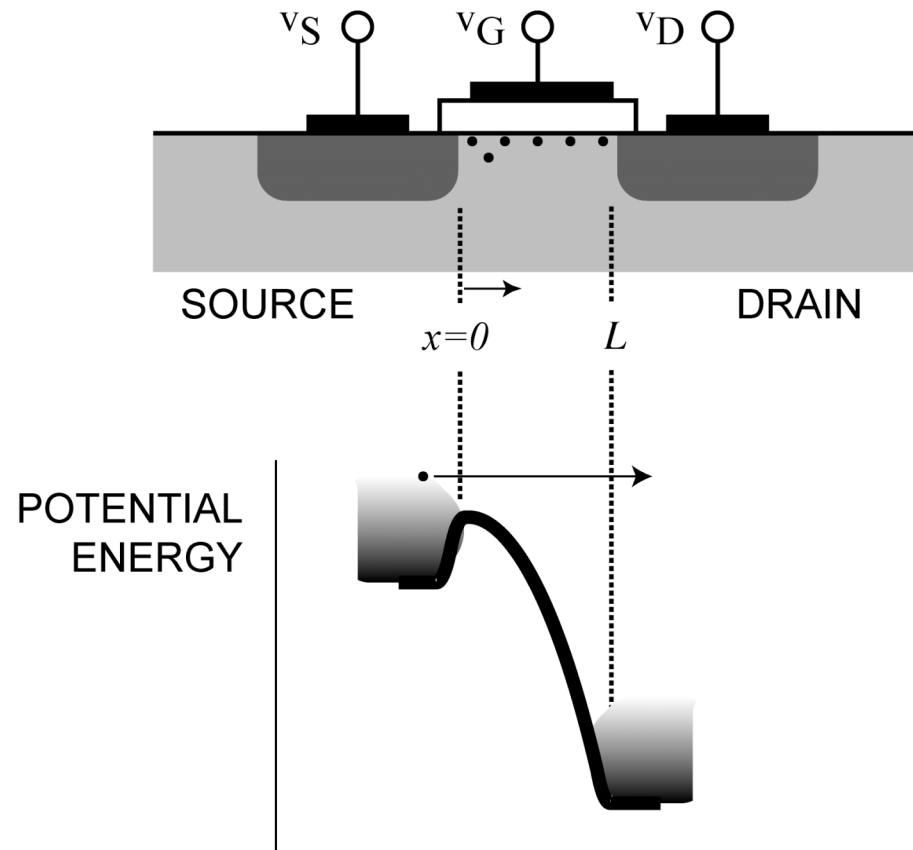
1. Carrier injection into channel
2. Low field motion modeled by mobility
3. Velocity saturated region

MOSFET Potential Energy ($L < \mu\text{m}$)



- Velocity saturated region is a greater fraction of channel
- Carriers still interact due to collisions

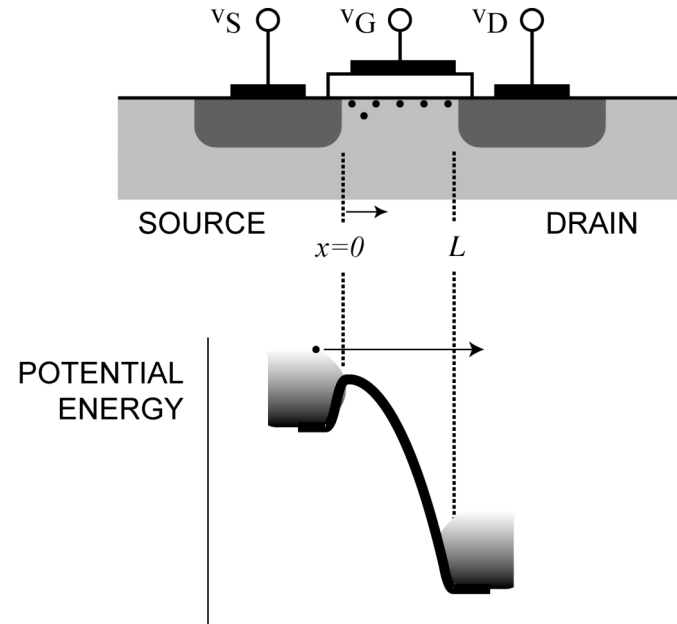
MOSFET Potential Energy ($L \ll \mu\text{m}$)



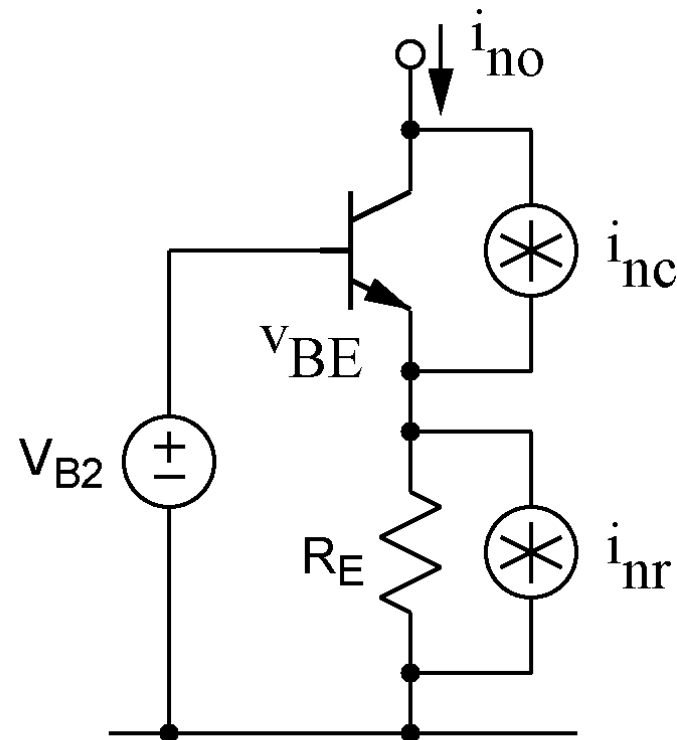
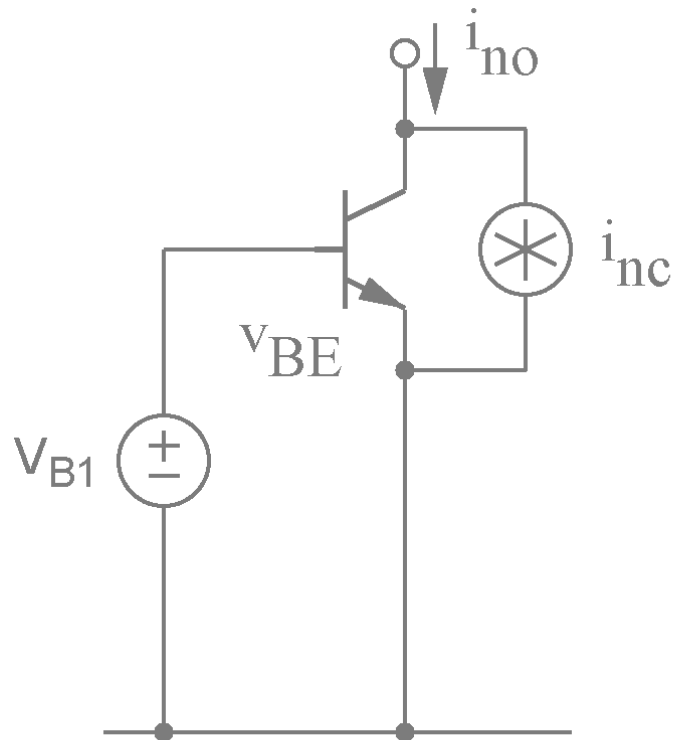
- Channel length $L \sim$ mean free path l_c
- "Ballistic": no interaction due to collisions
- No thermal equilibrium

$L < l_c$ "Breaking the Rules"

- $L < \text{mean free path } l_c$
- No thermal equilibrium
- No reason to expect any validity for a thermal noise / resistance model that assumed mobility and thermal equilibrium
- Behavior dominated by statistics of carrier injection at source
 - Shot noise! But not full shot noise:
 - Presence of injected carrier modifies potential profile; changes probability of injection

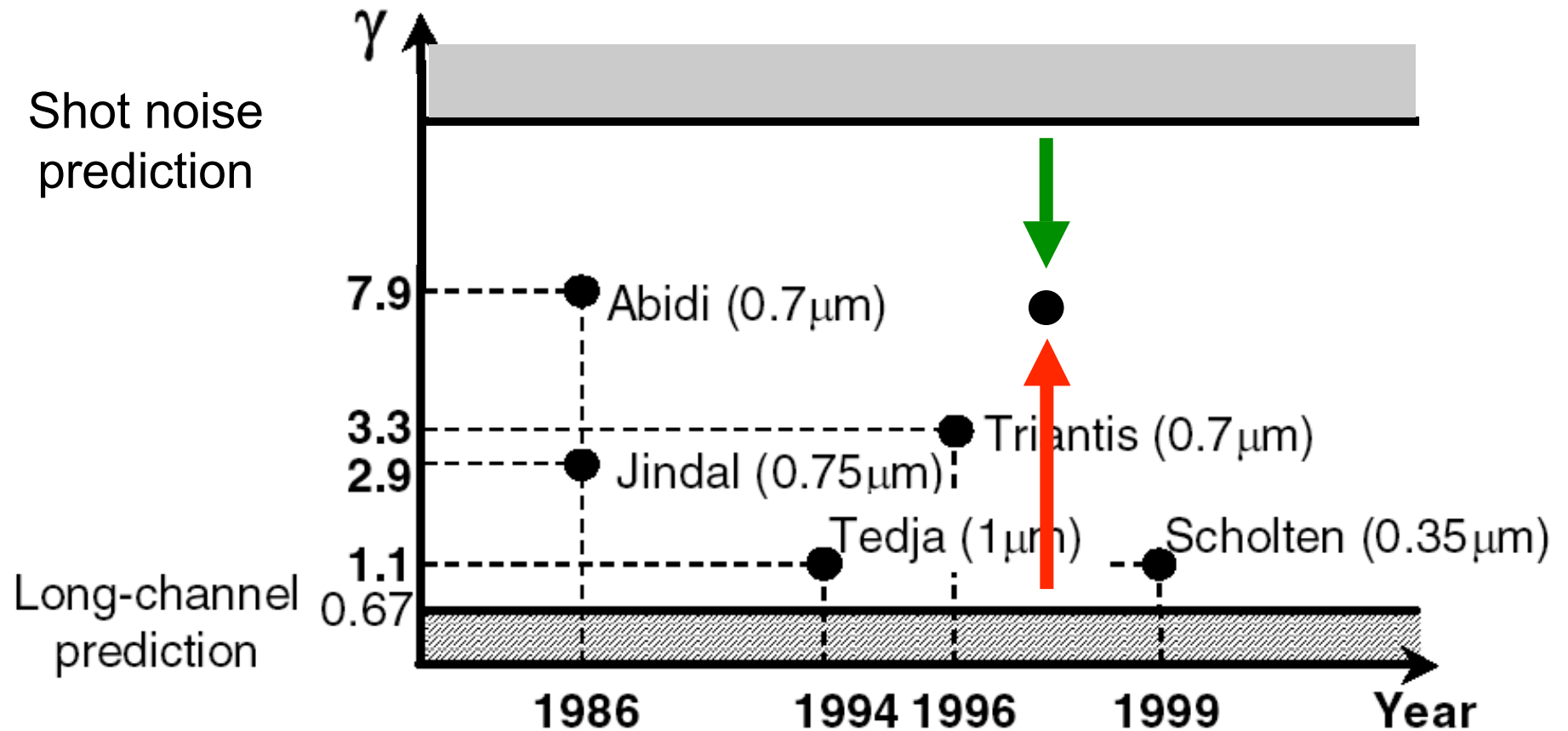


Analogy: Bipolar Transistor



- Output current noise i_{no} for isolated bipolar transistor is full shot noise i_{nc} of collector current
- With degeneration resistor: Not full shot noise:
- Voltage drop across R_E modifies v_{BE} ; feedback reduces variation in i_{no} due to i_{nc}

Submicron CMOS: Noise behavior



⇒ Don't interpret as γ increase

⇒ Interpret as shot noise suppression

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Jitter Example: Ring Oscillator

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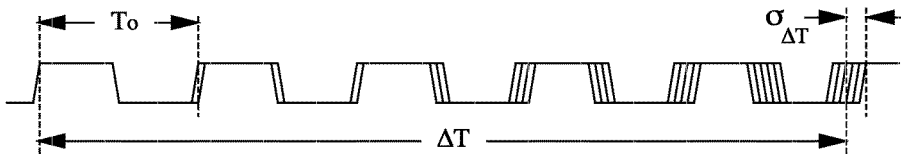
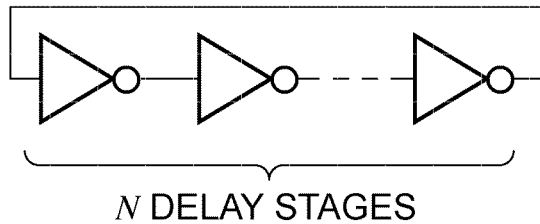
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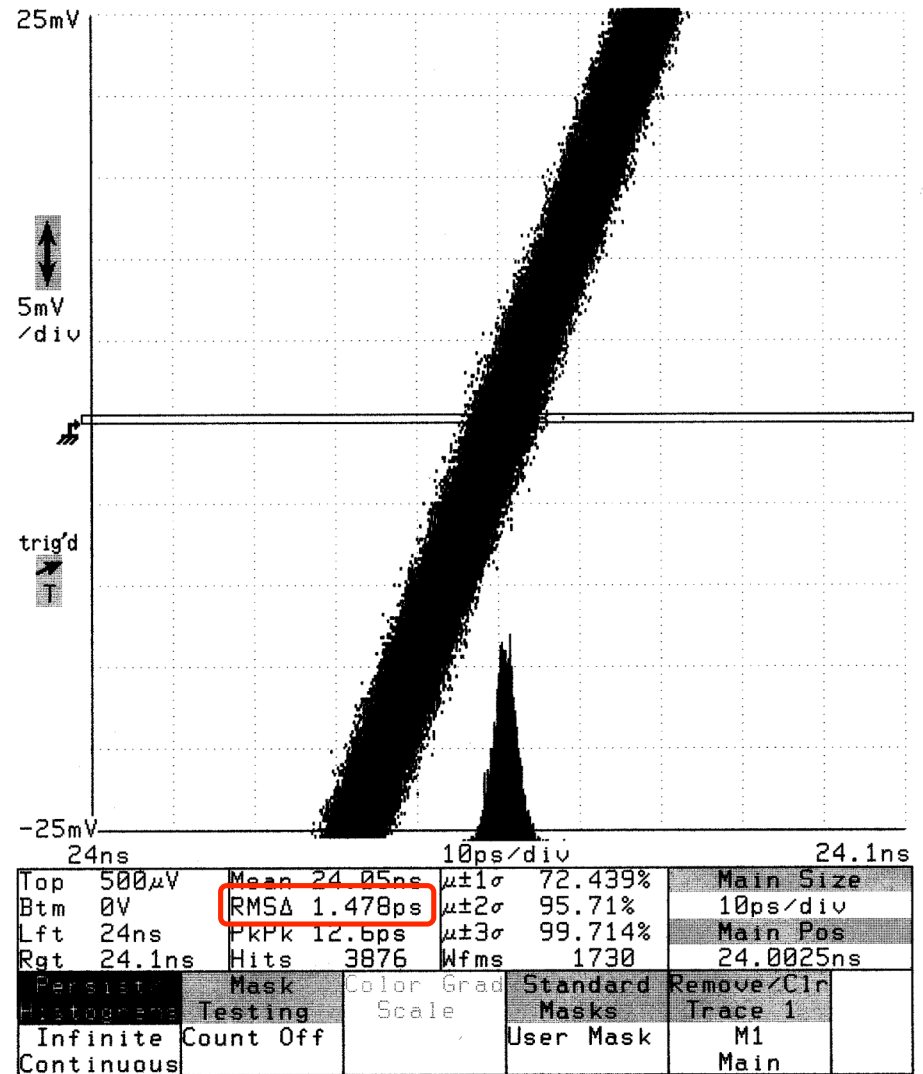
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Window FFTmag

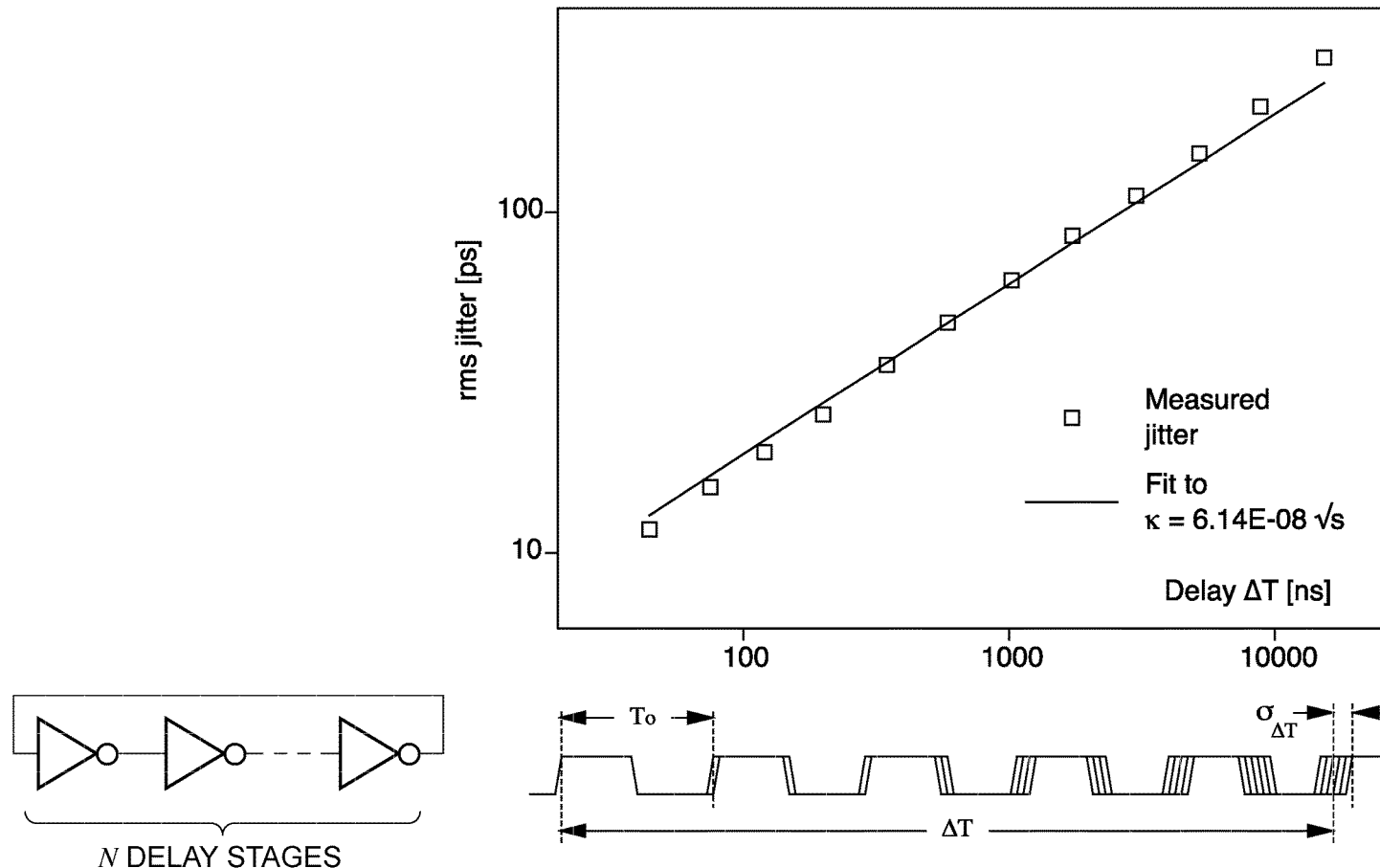
Def.Tra



- Time-domain noise (jitter) on clock transitions
- Characterized by standard deviation σ (ps rms)

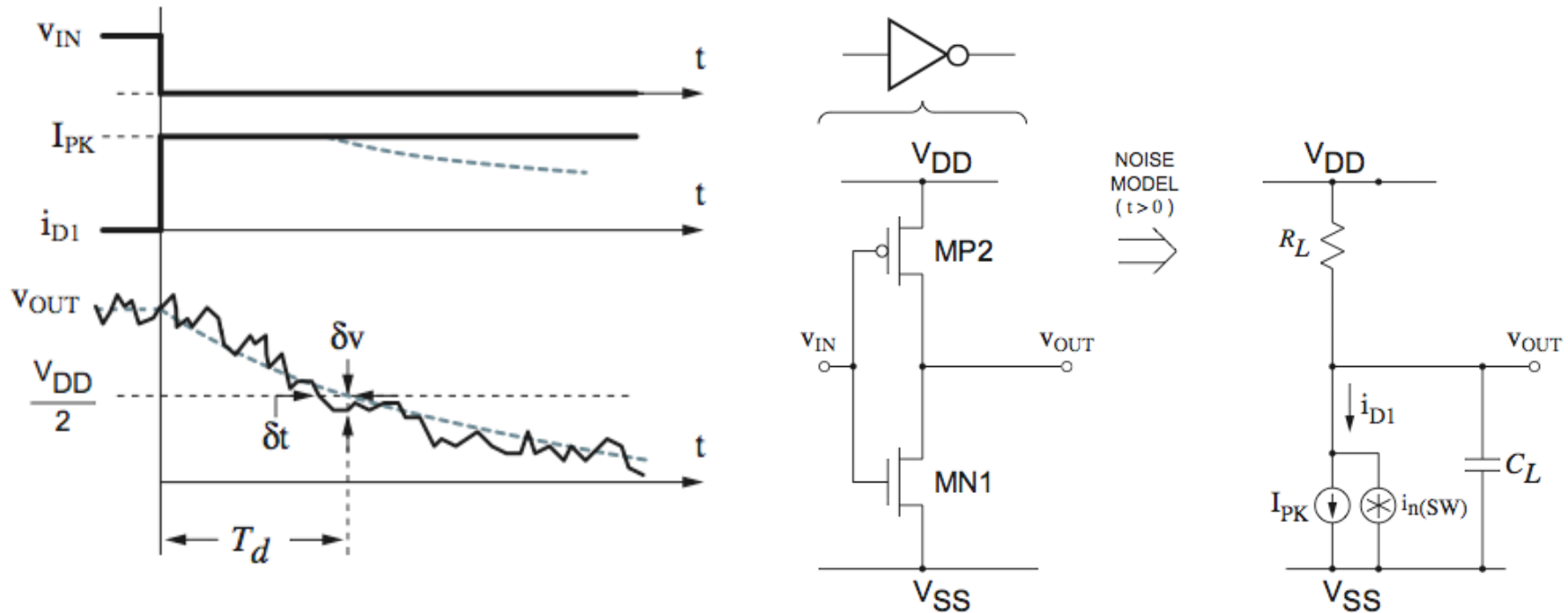


Jitter Example: Ring Oscillator



- Plot jitter vs. time interval ΔT
- Increases as square root: jitter $\sigma = \kappa \sqrt{\Delta T}$ delay
- κ frequency-independent figure-of-merit

Jitter at the Gate Delay Level



- MOSFET noise adds uncertainty to gate delay T_d
- Statistics of MOSFET noise can be related to oscillator figure-of-merit κ

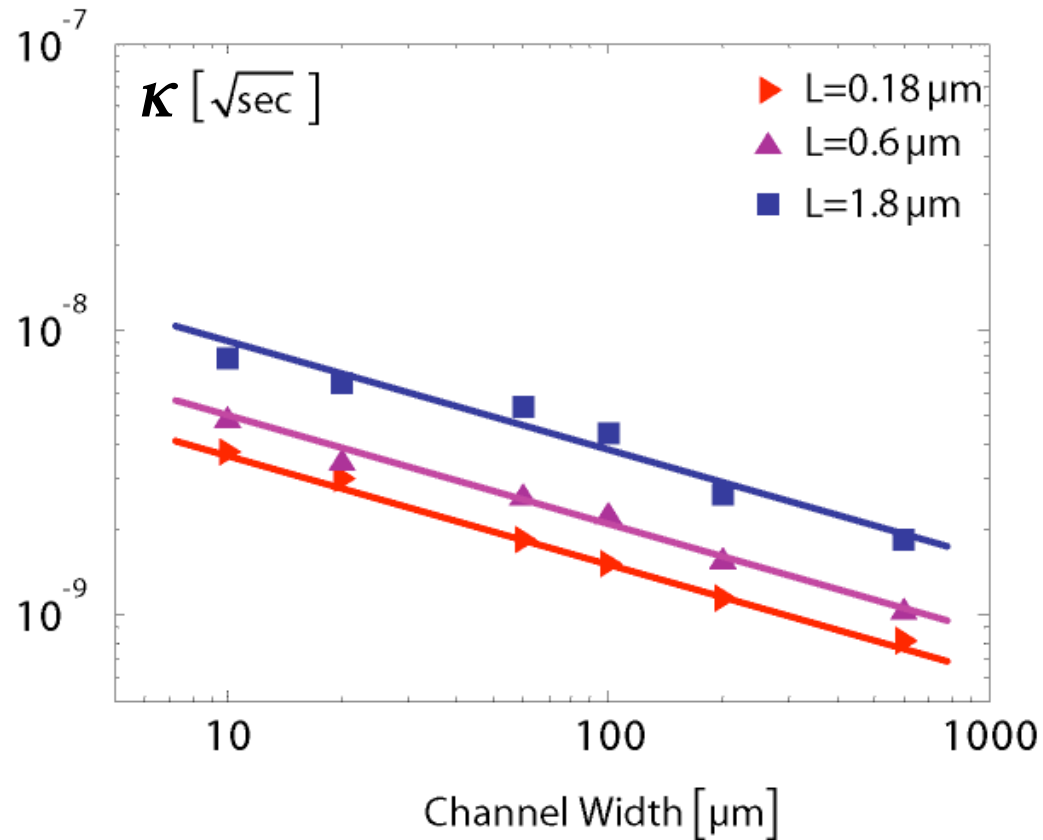
How to Improve Jitter?

- Burn more power
- Oscillator figure-of-merit κ of form

$$\kappa \propto \sqrt{\frac{kT}{POWER}}$$

- Derived from thermal noise model
- Intuitively, as oscillator power increases, random thermal energy is a smaller fraction of waveform

Oscillator Jitter κ vs. W



- Scales as predicted $\kappa \propto \sqrt{W}$

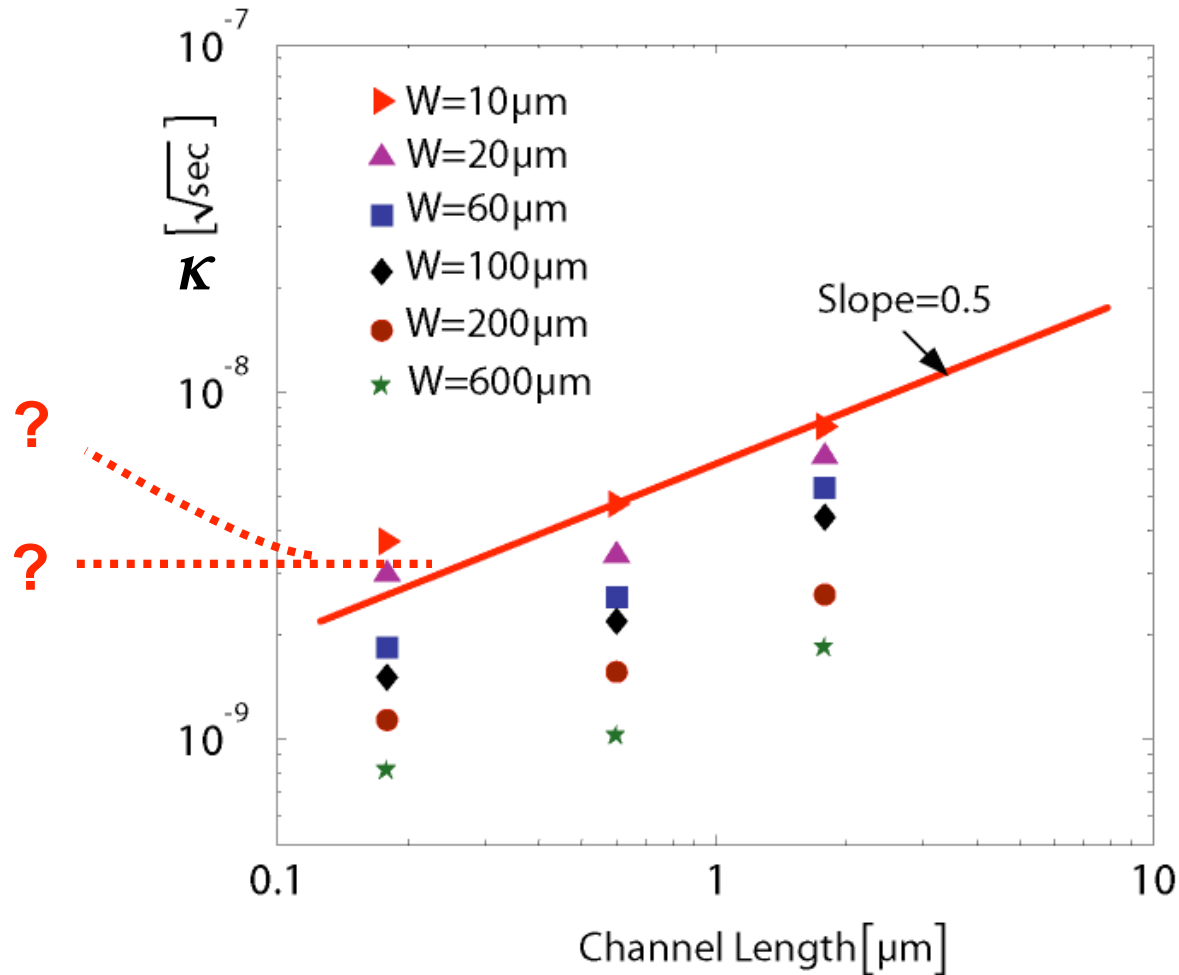
How to Improve Jitter?

- Burn more power
- Oscillator figure-of-merit κ of form

$$\kappa \propto \sqrt{\frac{kT}{POWER}}$$

- Derived from thermal noise model
- How does this behave as L shrinks?

Oscillator Jitter κ vs. L



- Deviation from predicted $\kappa \propto \sqrt{1/L}$ for $L < 1\mu\text{m}$
- Inflection or minimum?

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Learn from every outcome

DSM CMOS Conclusions

- **Survey the landscape**
 - Noise behavior changes for short L
- **Question assumptions**
 - Mobility model
- **Ask "what if" questions**
 - What if it's not a resistor?
- **Learn from every outcome**
 - Jitter example: Scaling may not provide benefits for analog as one might expect from long channel model

Design Drivers in DSM CMOS

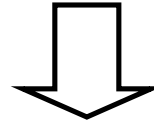
Explorer

**Environment: Decreasing ability
to predict analog performance
from simple assumptions / models**

Artist

Add Complexity

Digital



Judge

Eliminate Complexity

Analog

Warrior

Acknowledgments

- **WPI**
 - David Cyganski
 - Chengxin Liu
- **Analog Devices**
 - Mike Coln
 - Bob Adams
 - Larry DeVito
 - Colin Lyden
- **Carnegie Mellon**
 - David Ricketts
- **Columbia University**
 - Yannis Tsividis
- **Creativity Resources**
 - Roger von Oech

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